Consistent Discretization of a Class of Predefined-Time Stable Systems

Esteban Jiménez-Rodríguez * Rodrigo Aldana-López **
Juan D. Sánchez-Torres *** David Gómez-Gutiérrez ****†
Alexander G. Loukianov ‡

* Relativity6 Inc. Computer Science & Artificial Intelligence Lab,
Jalisco, 45040, México (e-mail: ejsjimezro@gmail.com)
** Department of Computer Science and Systems Engineering,
University of Zaragoza, Zaragoza, 50009 España, (e-mail: rodrigo.aldana.lopez@gmail.com)
*** Research Laboratory on Optimal Design, Devices and Advanced
Materials -OPTIMA-, Department of Mathematics and Physics, ITESO,
Jalisco, 45604 México (e-mail: dsanchez@iteso.mx)
**** Multi-agent autonomous systems lab, Intel Labs, Intel Tecnología de
México, Jalisco, 45019 México (e-mail: david.gomez.g@ieee.org)
† Tecnológico de Monterrey, Escuela de Ingeniería y Ciencias, Jalisco, 45138 México
‡ Department of Electrical Engineering, Cinvestav-Guadalajara, Jalisco,
45019 México (e-mail: louk@gdl.cinvestav.mx)

Abstract: As the main contribution, this document provides a consistent discretization of a
class of fixed-time stable systems, namely predefined-time stable systems. In the unperturbed
case, the proposed approach allows obtaining not only a consistent but exact discretization of the
considered class of predefined-time stable systems, whereas in the perturbed case, the consistent
discretization preserves the predefined-time stability property. All the results are validated
through simulations and compared with the conventional explicit Euler scheme, highlighting
the advantages of this proposal.

Keywords: Predefined-time stability, Discrete-time systems, Digital implementation, Stability
of nonlinear systems, Fixed-time stability

1. INTRODUCTION

Given a system with tunable parameters whose origin is
fixed-time stable, generally, it is not straightforward, and
sometimes it is even impossible to achieve any desired
upper bound of the settling time through the selection of
the parameters of the system (Jiménez-Rodríguez
et al., 2019, Example 1). To overcome this drawback, a
class of dynamical systems that exhibit the property of
predefined-time stability has been studied within the last
six years (Sánchez-Torres et al., 2018).

Predefined-time stability refers to the property that exhibits a particular class of fixed-time stable systems
whose solutions converge to the origin before an arbitrary user-prescribed time, which is assigned through an
appropriate selection of the tunable parameters of the system (Sánchez-Torres et al., 2018). Due to its remarkable
features, the study of the properties and applications of the predefined-time stability notion has attracted much
attention, mainly in continuous time (Aldana-López et al.,
2019; Aldana-López et al., 2019; Muñoz-Vázquez et al.,
2019; Sánchez-Torres et al., 2019).

The numerical simulation examples conducted in the
above works were done applying the conventional explicit
(forward) Euler discretization, with a tiny step size (1×10^{-4}
or 1×10^{-5} time units). However, the right side of the
ordinary differential equations exhibiting predefined-time
stability does not satisfy the Lipschitz condition at the
origin. In this case, the explicit Euler discretization scheme
does not guarantee that predefined-time stability property
will be preserved (Levant, 2013; Huber et al., 2016; Polyakov
et al., 2019). In other words, the discrete-time equation
solutions may be inconsistent with the solutions of the
continuous-time one.

In this sense, this work concerns with the discretization of a
class of predefined-time stable systems. The discretization
process uses the idea of (Polyakov et al., 2019, Figure 1.1),
but different from the contribution of the mentioned paper,
which achieves consistent discretizations of homogeneous
finite-time and practically fixed-time stable systems, the
proposed approach allows the consistent discretization of the
class of predefined-time stable systems proposed by
Aldana-López et al. (2019); Jiménez-Rodríguez et al.
(2019). Hence, this document provides a consistent
discretization of a class of fixed-time stable systems.

The rest of this paper is organized as follows: $K^1$ functions
and the considered class of predefined-time stable systems
are introduced in Section 2. Then, the exact discretization

of the considered class of predefined-time stable systems is developed in Section 3, in two different ways. Finally, Section 4 considers a consistent discretization of perturbed systems with discontinuous predefined-time control.

2. PRELIMINARIES

2.1 Class $K^1$ functions

The following definition was inspired on the comparison class $K$ functions (Kellett, 2014, Definition 1).

Definition 1. ($K^1$ functions). A scalar continuous function $\kappa : \mathbb{R}_{\geq 0} \rightarrow [0,1)$ is said to belong to class $K^1$, denoted as $\kappa \in K^1$, if it is strictly increasing, $\kappa(0) = 0$ and $\lim_{r \rightarrow \infty} \kappa(r) = 1$.

Remark 2. (Invertibility of class $K^1$ functions). Let $\kappa \in K^1$.

(i) $\kappa$ is injective since it is strictly increasing;

(ii) $\kappa$ is onto since it is continuous, $\kappa(0) = 0$ and $\lim_{r \rightarrow \infty} \kappa(r) = 1$.

Hence, $\kappa$ is bijective and its inverse exist.

Remark 3. (A connection with probability theory). Let $\kappa \in K^1$ be continuously differentiable on its domain. In this case, there exists a continuous function $\Phi : \mathbb{R}_{\geq 0} \rightarrow \mathbb{R}$ such that $\kappa(r) = \int_0^r \Phi(z)dz$. Moreover, since $\lim_{r \rightarrow \infty} \kappa(r) = 1$, the function $\Phi$ is required to satisfy $\int_0^\infty \Phi(z)dz = 1$, i.e., functions $\Phi$ and $\kappa$ can be viewed as probability density functions and cumulative distribution functions, respectively, of positive random variables.

Example 4. Examples of class $K^1$ functions are:

- $\kappa(r) = \frac{2}{\pi} \arctan(ar)$, with $a > 0$;
- $\kappa(r) = \frac{1}{r^{\rho_1}}$, with $a > 0$;
- $\kappa(r) = P(a,r)$, with $a > 0$, where $P(a,r) = \frac{\gamma(a,r)}{\Gamma(a)}$ is the regularized Incomplete Gamma Function, $\gamma(a,r) = \int_0^r t^{a-1}e^{-t}dt$ is the Incomplete Gamma Function and $\Gamma(a) = \int_0^\infty t^{a-1}e^{-t}dt$ is the Gamma Function (Abramowitz and Stegun, 1965);
- $\kappa(r) = I(a_1,a_2,r)$, with $a_1, a_2 > 0$, where $I(a_1,a_2,r) = \frac{b(a_1,a_2,r)}{B(a_1,a_2)}$ is the regularized Incomplete Beta Function, $b(a_1,a_2,r) = \int_0^r t^{a_1-1}(1-t)^{a_2-1}dt$ is the Incomplete Gamma Function and $B(a_1,a_2) = \int_0^1 t^{a_1-1}(1-t)^{a_2-1}dt$ is the Beta Function (Abramowitz and Stegun, 1965).

2.2 A class of predefined-time stable systems

Consider (the scalar form of) the class of systems introduced by Aldana-López et al. (2019) and Jiménez-Rodríguez et al. (2019)

$$\dot{x} = -\frac{1}{\rho_1(1-\rho_2)} \frac{\kappa([x])^{\rho_2}}{\kappa([x])} \text{sign}(x), \ x(0) = x^0,$$  \hspace{1cm} (1)

where $x : \mathbb{R}_{\geq 0} \rightarrow \mathbb{R}$ is the state of the system, $\rho_1 > 0$ and $0 \leq \rho_2 < 1$ are tunable parameters, $\kappa \in K^1$ is continuously differentiable, and $\text{sign}(\cdot)$ stands for the signum function given by

$$\text{sign}(x) \in \begin{cases} 
\{1\} & \text{if } x > 0 \\
\{-1\} & \text{if } x < 0 \\
\{-1,1\} & \text{if } x = 0.
\end{cases}$$

From the theory of ordinary differential equations, (1) is a separable first-order equation, whose solution $x(t)$ satisfies

$$\int_0^{x(t)} (1-\rho_2) \frac{\kappa([\xi])^{\rho_2}}{\kappa([\xi])} \text{sign}(\xi) d\xi = -\int_0^t \frac{1}{\rho_1} dr.$$

Applying the change of variable $\xi \rightarrow \kappa([\xi])^{1-\rho_2}\text{sign}(\xi)$ in the left side integral and integrating both sides of the above equation, it yields

$$\kappa([x(t)])^{1-\rho_2} = \kappa([x^0])^{1-\rho_2} - \frac{1}{\rho_1},$$

for $\kappa([x^0])^{1-\rho_2} - \frac{1}{\rho_1} \geq 0$. From the above, the solution of (1) is

$$x(t) = \begin{cases} 
\kappa^{-1} \left( \kappa([x^0])^{1-\rho_2} - \frac{1}{\rho_1} \right)^{1-\rho_2} \text{sign}(x^0) & \text{if } 0 \leq t \leq \rho_1 \kappa([x^0])^{1-\rho_2} \\
0 & \text{if } t > \rho_1 \kappa([x^0])^{1-\rho_2},
\end{cases}$$  \hspace{1cm} (2)

where $\kappa^{-1}$ stands for the inverse of the function $\kappa \in K$.

Hence, one can conclude that the origin $x = 0$ of (1) is finite-time stable and the settling-time function is $T(x^0) = \rho_1 \kappa([x^0])^{1-\rho_2}$. But the settling-time function $T(x^0) = \rho_1 \kappa([x^0])^{1-\rho_2} < \rho_1$ is bounded, then the origin $x = 0$ of (1) is, in fact, fixed-time stable.

Moreover, for any desired upper bound of the settling-time function $T_c$, there exists an appropriate selection of the parameter $\rho_1 = T_c$ such that the convergence time will always be less than $T_c$ for any initial condition $x^0$. This property is known as predefined-time stability (see Appendix A).

Remark 5. It is worth noticing that the right-hand side of (1) is continuous and non-Lipschitz if $0 < \rho_2 < 1$, and discontinuous if $\rho_2 = 0$. In any case, all the above derivation remains true.

3. EXACT DISCRETIZATION OF THE CONSIDERED CLASS OF PREDEFINED-TIME STABLE SYSTEMS

In this section, we develop a discrete-time system whose solutions are equal to the solution of (1) in the sampling instants. Hence, we provide the exact form of simulating the class of predefined-time stable systems represented by (1).

3.1 Direct construction

Consider the continuous-time system (1) and let $x_k = x(kh)$, where $h > 0$ is the sampling step size, and $k \in \mathbb{N}$.

To obtain the exact discretization of system (1), one can apply the same procedure as in Subsection 2.2, integrating the left side between the samples $x_k$ and $x_{k+1}$, and the right side between $kh$ and $(k+1)h$, respectively, obtaining

$$\int_{x_k}^{x_{k+1}} (1-\rho_2) \frac{\kappa([\xi])^{\rho_2}}{\kappa([\xi])^{\rho_2}} \text{sign}(\xi) d\xi = -\frac{h}{\rho_1}.$$

Recall the transformation

$$y_{\rho_2} : \mathbb{R} \rightarrow (-1,1), \ x \mapsto y_{\rho_2}(x) = \kappa([x])^{1-\rho_2}\text{sign}(x)$$  \hspace{1cm} (4)
used in Subsection 2.2 to find the solutions of (1). This transformation is invertible, and its inverse \( y^{-1}_{\rho_2} : (−1,1) \to \mathbb{R} \) is given by

\[
y^{-1}_{\rho_2}(x) = \kappa^{-1} \left( |x|^{1−\rho_2} \right) \cdot \text{sign}(x).
\]

Using (4) in the left side integral of (3), one obtains

\[
\int_{k(x_k)}^{\kappa(|x_{k+1}|)^{1−\rho_2} \cdot \text{sign}(x_{k+1})} \cdot \text{sign}(y) \, dy = -\int_{k+1}^{k} \frac{1}{\rho_1} \, dt.
\]

Hence, integrating both sides of the above expression and having in mind that \( \text{sign}(x_{k+1}) = \text{sign}(x_k) \) for the first-order system (1), the above equation turns into

\[
\kappa(|x_{k+1}|)^{1−\rho_2} \cdot \text{sign}(x_{k+1}) = \kappa(|x_k|)^{1−\rho_2} \cdot \text{sign}(x_k) - \frac{h}{\rho_1} \cdot \text{sign}(x_{k+1}). \tag{5}
\]

Finally, let \( \alpha \in \mathbb{R} \) and consider the discontinuous function \( \theta_{\rho_2, \alpha} : \mathbb{R} \to (−\alpha − 1, \alpha + 1) \) defined by

\[
\theta_{\rho_2, \alpha}(x) = (\kappa(|x|)^{1−\rho_2} + \alpha) \cdot \text{sign}(x).
\]

The inverse of \( \theta_{\rho_2, \alpha} \) is the continuous function \( \theta_{\rho_2, \alpha}^{-1} : (−\alpha − 1, \alpha + 1) \to \mathbb{R} \) given by

\[
\theta_{\rho_2, \alpha}^{-1}(x) = \kappa^{-1} \left( (\max\{|x| - \alpha, 0\})^{1−\rho_2} \right) \cdot \text{sign}(x). \tag{6}
\]

Thus, replacing (6) in (5), it yields

\[
x_{k+1} = \theta_{\rho_2, h/\rho_1}^{-1} (\kappa(|x_k|)^{1−\rho_2} \cdot \text{sign}(x_k)),
\]

\[
= \kappa^{-1} \left( \max \left\{ \kappa(|x_k|)^{1−\rho_2} - \frac{h}{\rho_1} \cdot 0, 0 \right\}^{1−\rho_2} \right) \cdot \text{sign}(x_k), \tag{7}
\]

with \( x_0 = x^0 \).

This following theorem summarizes the previous constructive result:

**Theorem 6.** The solution of the discrete-time system (7) is equivalent to the solution (2) of the continuous-time system (1) at the sample instants. This is, \( x_k = x(kh) \), for every \( k \in \mathbb{N} \).

Theorem 6 has the following attractive corollary:

**Corollary 7.** For any \( x^0 \in \mathbb{R} \), the solution of the discrete-time system (7) satisfies \( x_k = 0 \), for \( k \geq \left\lceil \frac{x_0}{h} \right\rceil \), where \( \lceil \cdot \rceil \) stands for the ceiling function.

**Proof.** The solution of the continuous-time system (1) satisfies \( x(t) = 0 \), for \( t \geq \rho_1 \kappa \left( |x^0| \right)^{1−\rho_2} \).

Moreover, from Theorem 6, we can establish for the solution of the discrete-time system (7) that

\[
x_k = x(kh) = 0, \text{ for } kh \geq \rho_1 \kappa \left( |x^0| \right)^{1−\rho_2},
\]

which also holds for \( kh \geq \rho_1 \), given that \( \kappa \in K \).

Hence, the result follows. □

**Example 8.** Consider system (1) with the particular selections of \( \kappa(r) = \frac{2}{\pi} \cdot \arctan(r) \), \( \rho_1 = 1 \), \( \rho_2 = 0.5 \) and \( x^0 = 10 \).

Figures 1 and 2 present the comparison of the exact discretization (7) with the conventional explicit Euler discretization for two different time steps. In both cases, the solution of the discrete-time system (7) accurately recovers the exact solution (2) at the sampling instants.

For \( h = 0.02 \) (Fig. 1), one can note small differences between the exact solution and the explicit Euler discretization, whereas, for \( h = 0.0651318636497 \) (Fig. 2), the explicit Euler discretization becomes unstable.

**Remark 9.** If the explicit or forward Euler procedure with time step \( h > 0 \) were used to simulate system (8), it would lead to oscillations with amplitude \( O(h) \) in the...
neighborhood of the origin $w = 0$ (Drakunov and Utkin, 1989; Utkin, 1994), and these oscillations would also be translated into oscillations in the original variable $x$. Those mentioned oscillations, known as numerical chattering, are an undesired effect of an inadequate discretization, since the solution (2) does not oscillate.

Using the implicit or backward Euler method (as in (Drakunov and Utkin, 1989; Utkin, 1994)) with time step $h > 0$ for the discretization of the transformed system (8), it yields

$$w_{k+1} = w_k - \frac{h}{\rho_1} \text{sign}(w_{k+1}),$$

where $w_k = y_{pk}(x_k)$ and $x_k = x(kh)$. Rewriting the above in terms of $x_k$, one obtains (3), as before.

4. CONSISTENT DISCRETIZATION OF PREDEFINED-TIME CONTROL OF PERTURBED SYSTEMS

4.1 Continuous-time predefined-time stabilization of first order perturbed systems

Consider the perturbed control system

$$\dot{x} = u + \Delta(t, x), \quad x(0) = x^0,$$

where $x : \mathbb{R}_{\geq 0} \to \mathbb{R}$ is the state of the system, $u \in \mathbb{R}$ is the control input signal, and $\Delta : \mathbb{R}_{\geq 0} \times \mathbb{R} \to \mathbb{R}$ is an unknown perturbation term which is assumed to be bounded of the form $\sup_{t,x} |\Delta(t, x)| \leq \delta$, with $\delta \in \mathbb{R}$ plus known.

It is well known that, in continuous time, the perturbation term $\Delta(t, x)$ can only be entirely rejected by a discontinuous control term, since no conditions of smoothness, Lipschitz continuity neither continuity are assumed (Jiménez-Rodríguez et al., 2019).

According to the above, a suitable feedback predefined-time controller design is

$$u = - \left( \frac{1}{\rho_1} + \rho_3 \kappa'(0) \right) \frac{1}{\kappa'(|x|)} \text{sign}(x),$$

where $\rho_1 > 0$, $\rho_3 \geq \delta$, and $\kappa \in \mathcal{K}_1$ is continuously differentiable and such that $\kappa'(0) := \kappa'(0^+) \geq \kappa'(r)$ for all $r \in \mathbb{R}_{\geq 0}$. Note that (10) is an augmented gain version of the right side of (1) with $\rho_2 = 0$.

To verify the predefined-time convergence of the closed-loop system (9)-(10), note that

$$\frac{d|x|}{dt} = - \left( \frac{1}{\rho_1} + \rho_3 \kappa'(0) \right) \frac{1}{\kappa'(|x|)} \text{sign}(x) + \Delta(t, x)\text{sign}(x)$$

$$\leq - \frac{1}{\rho_1 \kappa'(|x|)} - \rho_3 \frac{\kappa'(0)}{\kappa'(|x|)} + |\Delta(t, x)|$$

$$\leq - \frac{1}{\rho_1 \kappa'(|x|)} - (\rho_3 - \delta)$$

$$\leq - \frac{1}{\rho_1 \kappa'(|x|)},$$

for $x \neq 0$. Reasoning as in Subsection 2.2, and using the comparison lemma in the above differential inequality, one can easily conclude that $|x(t)| = x(t) = 0$ for $t \geq \rho_1$.

4.2 Consistent discretization/simulation

Consider transformation (4), with $\rho_2 = 0$, given by

$$z(t) = y_0(x(t)) = \kappa(|x(t)|)\text{sign}(x(t)).$$

Its derivative along the trajectories of the closed-loop system (9)-(10) is

$$\dot{z} = -\beta \text{sign}(z) + f(t, x), \quad z(0) = y_0(x^0),$$

where $\beta := \frac{1}{\rho_1} + \rho_3 \kappa'(0)$, and $f(t, x) = \kappa'(|x|)\Delta(t, x)$ is the perturbation term in the transformed coordinate $z$, which complies to $\sup_{t,x} f(t, x) \leq \kappa'(0)\delta$.

Again, using the implicit Euler method with time step $h > 0$ for the discretization of the system (11), it yields

$$z_{k+1} = z_k - h \beta \text{sign}(z_{k+1}) + f_k,$$

where $f_k = f(kh, x_k)$ is the perturbation term at the sample instants.

Hence, replacing the function $\theta_{0,h\beta}^{-1}$ (6) into (12), one obtains

$$x_{{k+1}} = \kappa^{-1} \left( \max \left\{ \kappa(|x_k|)\text{sign}(x_k) + f_k - \delta, 0 \right\} \right) \times \text{sign}(\kappa(|x_k|)\text{sign}(x_k) + f_k),$$

with $x_0 = x^0$.

The discrete-time system (13) provides a way of simulating the closed-loop continuous-time system (9)-(10). Moreover, it preserves the predefined-time stability behavior, as stated in the following proposition:

**Proposition 10.** For any $x^0 \in \mathbb{R}$, the solution of the discrete-time system (13) satisfies $x_k = 0$, for $k \geq \left[ \frac{x^0}{h} \right]$.

**Proof.** Taking absolute value in both sides of (13) and using the triangle inequality one gets,

$$|x_{k+1}| = \kappa^{-1} \left( \max \left\{ \kappa(|x_k|)\text{sign}(x_k) + h f_k, 0 \right\} \right) \leq \kappa^{-1} \left( \max \left\{ \kappa(|x_k|) + h f_k - \delta, 0 \right\} \right) \leq \kappa^{-1} \left( \max \left\{ \kappa(|x_k|) + h f_k - \delta, 0 \right\} \right),$$

On the other hand, from $\beta := \frac{1}{\rho_1} + \rho_3 \kappa'(0)$, $\rho_3 \geq \delta$, $|f_k| \leq \kappa'(0)\delta$ and the above, the inequality

$$|x_{k+1}| \leq \kappa^{-1} \left( \max \left\{ \kappa(|x_k|) + h f_k - \delta, 0 \right\} \right)$$

is obtained.

Finally, noticing that the right side of the above inequality is equal to the absolute value of the right side of (7) with $\rho_2 = 0$, and that it is a non-decreasing function of $x_k$, we get

$$|x_k| \leq \alpha_k \geq \left[ \frac{\rho_1}{h} \right]$$

applying Lemma 13.

**Example 11.** Consider the closed-loop system (9)-(10) with the particular selection of $\kappa(r) = \frac{2}{\pi} \arctan(r)$, $\rho_1 = 1$, $\rho_3 = 1.1$ and $x^0 = 10$. It is assumed that $\Delta(t, x) = \sin(10t)$.

Figure 3 present the comparison of the consistent discretization (13) with the conventional explicit Euler discretization for $h = 0.05$. One can see that $x_k = 0$ for $k \geq \left[ \frac{x^0}{h} \right]$ for the consistent discretization (13), whereas the explicit Euler discretization induce undesired oscillations around the origin.
5. CONCLUSION

This paper presented the development of a consistent discretization for the class of predefined-time stable systems introduced in Aldana-López et al. (2019); Jiménez-Rodríguez et al. (2019).

The proposed approach allowed the exact discretization of the considered class of systems when no perturbations were assumed. In the perturbed case, the developed consistent discretization preserved the predefined-time stability property.

All the results were confirmed through numerical simulations and compared with the conventional explicit Euler scheme. Even with relatively large time steps, the proposed discretization worked well, as expected, whereas the explicit Euler discretization produced unstable oscillations.

ACKNOWLEDGEMENTS

Esteban Jiménez acknowledges to CONACYT–México for the D.Sc. scholarship number 481467 and the project 252405.

REFERENCES


Appendix A. PREDEFINED-TIME STABILITY

Predefined-time stability refers to the property that exhibits a particular class of fixed-time stable systems with tunable parameters, for which an upper bound of the settling-time function can be arbitrarily chosen through a suitable selection of the parameters (Jiménez-Rodríguez et al., 2019, Definition 2).

This notion is formally defined considering an autonomous system of the form

\[ \dot{x} = f(x; \rho), \quad x(0) = x_0, \]  

(A.1)

with \( x : \mathbb{R}^n \to \mathbb{R}^n \) the system state, \( \rho \in \mathbb{R}^l \) the tunable parameters of (A.1), and \( f : \mathbb{R}^n \to \mathbb{R}^n \) a nonlinear function.

For the origin \( x = 0 \) to be a finite-,

, in particular, a predefined-time stable equilibrium of (A.1), the function \( f \)
must be a non Lipschitz (maybe discontinuous) function of $x$. Then, $f$ is assumed to be such that the solutions of (A.1) exist and are unique in the sense of Filippov (Filippov, 1988).

**Definition 12.** The origin of (A.1) is said to be **predefined-time stable** if it is fixed-time stable and for any $T_c \in \mathbb{R}_+$, there exists some $\rho \in \mathbb{R}$ such that the settling-time function of (A.1) satisfies
\[
\sup_{x_0 \in \mathbb{R}^n} T(x_0) \leq T_c.
\]

**Appendix B. A COMPARISON LEMMA FOR DISCRETE-TIME SYSTEMS**

The following is a particular case of (Bitsoris and Gravalou, 1995, Proposition 1).

**Lemma 13.** Let $u_k$ be the solution of
\[
u_{k+1} = f(u_k), \quad u_0 = u^0, \quad u \in \mathbb{R},
\]
where $f : \mathbb{R} \rightarrow \mathbb{R}$ is continuous and non-decreasing, and let $v_k$ be such that
\[
v_{k+1} \leq f(v_k), \quad v_0 \leq u^0, \quad v \in \mathbb{R}.
\]

Then, $v_k \leq u_k$, for all $k \in \mathbb{N}$.

**Proof.** We proceed by induction. The base case follows from the hypothesis
\[
v_0 \leq u^0 = u_0.
\]
Now, assume that $v_k \leq u_k$. Then,
\[
v_{k+1} \leq f(v_k) \leq f(u_k) = u_{k+1},
\]
since $f$ is non-decreasing. The result follows. ■