

Controllability of linear delay systems and of their sampled versions

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Abstract: This paper focuses on the controllability preservation through sampling of linear time-delay systems. We make use of a module theoretic framework acting as a unifying one for most of the existing delay system controllability notions. The controllability properties are envisioned through ring theoretic properties. Some illustrative examples complete the presentation.

Keywords: Delay systems, Controllability, Sampling, Module theoretic setting, π -freeness.

1. INTRODUCTION

Delay systems, e.g. systems modeled by differential-difference equations represent the simplest class of infinite-dimensional processes. They naturally arise in technological systems, where the delay is quite often situated in the input or the output. This infinite dimensional character is merely lost through discretization.

Contrarily to the case of finite-dimensional systems, where the controllability notion is unique, there has been numerous, and seemingly unrelated controllability generalizations for delay systems. More specifically, notions have been elaborated either from a functional analytic approach (see, e.g., Bartosiewicz (1984); Manitius and Triggiani (1978); Yamamoto (1989)) or from a more algebraic viewpoint (see, e.g., with a similar formalism as used here (Bourlès and Marinescu (2011); Fabiańska and Quadrat (2007); Mounier and Rudolph (2003); Rudolph et al. (2003)), with systems over rings (Loiseau (2000); Sontag (1976)), with distributional rings Bourlès and Oberst (2015), Vettori and Zampieri (2002)), with dual trajectory modules over an operator ring (see, for instance, Gluesing-Luerssen (2002); Pillai and Shankar (1999)). There has thus also been many bridging notions between continuous delay systems and sampled data systems (Fridman (2014a,b); Louisell (2001)).

In Fliess and Mounier (1998), an algebraic setting was introduced, based on module theoretic properties, which yielded a unifying framework within which most of these notions could be compared. Having the previous facts in mind, it is natural to ask *which controllability property can be preserved under sampling*.

The main features of our algebraic framework can be summarized as follows:

- The *intrinsic character* of the various notions and definitions.

- Varying the coefficients ring and the module properties yield a *great flexibility* in defining structural notions.
- A framework enabling the *unification of most of the existing structural notions* of the delay system's literature (see Fliess and Mounier (1998)).
- The *constructive character* of the structural notions, as demonstrated by the numerous computer algebra implementations (see, e.g., Chyzak et al. (2005); Quadrat and Robertz (2014)).
- The *independence* of the obtained structural results from often tricky *existence and uniqueness* prerequisites, as is customary for infinite dimensional systems through functional analytic techniques.
- The *same framework* can be used both for *finite- and infinite-dimensional linear systems*.

The sampling we shall use are classical ones, and we will not explicitly study the effect of sampled data control on a continuous system.

The aim of this contribution is to investigate whether controllability notions can be preserved through sampling.

The remaining of the paper is organized as follows: In Section 2, general algebraic definitions are given, and the case of finite dimensional and delay systems are given in Section 3. Discretization and interplay between controllability and sampling are studied in Sections 4 and 5. Finally, some concluding remarks end the paper.

2. PREREQUISITES AND R -SYSTEMS

Most of the commutative algebra we use may be found in standard textbooks, such as Eisenbud (1995); Lang (71); Rotman (1979).

In particular, the basic definitions of module, ideal, free module, . . . , can be found for example in pages 11 to 17 of Eisenbud (1995). The notion of torsion freeness is in (Lang, 71, p. 388) or (Rotman, 1979, p. 224). For projectivity, see,

e.g., (Eisenbud, 1995, A3.2 p. 615) or (Rotman, 1979, Th. 3.11 and 3.14, p. 62–63).

2.1 Preliminary definitions

All the rings and algebras we shall consider are commutative, with unity, and without zero divisors.

Definition 2.1. Let R be a ring. An R -linear system, or an R -system, or a (linear) system over R , Σ is a finitely generated R -module. \square

In other words, an R -system is an R -module with finite free presentation (see, e.g., (Eisenbud, 1995, p.17), (Rotman, 1979, p. 90)):

$$E \xrightarrow{\phi} W \longrightarrow \Sigma \longrightarrow 0.$$

A presentation matrix P_Σ of Σ is one associated with ϕ .

NOTATION. We denote by $[\xi]$ the R -submodule spanned by a subset ξ of Σ .

Definition 2.2. An R -linear dynamics, or an R -dynamics, or a (linear) dynamics over R , is an R -system equipped with an input $\mathbf{u} = (u_1, \dots, u_m)$, i.e., a finite subset of Σ , such that the quotient module $\Sigma/[\mathbf{u}]$ is torsion. The input \mathbf{u} is said to be independent if $[\mathbf{u}]$ is a free R -module of rank m . \square

Definition 2.3. An output $\mathbf{y} = (y_1, \dots, y_p)$ is a finite subset of Σ . An R -system equipped with an input and an output is called an input-output R -system. \square

2.2 General controllability notions

Varying the rings under consideration through suitable tensor products yields various controllability notions.

Definition 2.4. Let A be an R -algebra. The R -system Σ is said to be A -torsion free controllable (resp. A -reflexive controllable, A -projective controllable, A -free controllable) if the A -module $A \otimes_R \Sigma$ is torsion free (resp. projective, free). \square

Elementary homological algebra Rotman (1979) yields:

Proposition 2.1. The A -free controllability implies the A -projective controllability, which implies the A -reflexive controllability, which implies the A -torsion free controllability. \square

Localisation (e.g. inversion of some elements) preserves module properties in the following sense (see, e.g., (Rotman, 1979, 4.81, p. 198)):

Proposition 2.2. If $S \subseteq R$ is multiplicative, then localisation $\Sigma \mapsto S^{-1}\Sigma = S^{-1}R \otimes_R \Sigma$ defines an exact functor $\mathbf{Mod}_R \rightarrow \mathbf{Mod}_{S^{-1}R}$. Thus if Σ is free (resp. projective), then $S^{-1}\Sigma$ is free (resp. projective). \square

The following result is borrowed from (Rowen, 1991, Proposition 2.12.17, p. 233):

Proposition 2.3. Let Σ be an R -system, A an R -algebra, and S a multiplicative part of A such that Σ is $S^{-1}R$ -free controllable. Then, there exists an element π in S such that Σ is $R[\pi^{-1}]$ -free controllable. \square

Definition 2.5. Let Σ be an R -system, A an R -algebra, and S a multiplicative part of A containing an element π such that Σ is $R[\pi^{-1}]$ -free controllable. Such a system Σ will be called π -free. \square

We have the following direct consequence:

Corollary 2.1. Let Σ be an R -torsion free controllable R -system and S a multiplicative part of R such that $S^{-1}R$ is a principal ideal ring. Then, there exists $\pi \in S$ such that Σ is $R[\pi^{-1}]$ -free controllable and Σ is π -free. \square

Remark 2.1. Note that the previous definitions allow for an extreme flexibility in the choice of the structural notions to be considered. First, three distinct algebraic notions are underlined (torsion freeness, projectivity and freeness) and last, but not least, the base change, through extension of scalars, yields a vast number of notions. \square

Let us end this subsection by another notion, the spectral controllability, very useful in the infinite-dimensional case. The following definition (given in Woittennek and Mounier (2010)) extends previous ones (see, e.g., Kirillova and Churakova (1967), Asmykovich and Marchenko (1976), Bartosiewicz (1984), Rocha and Willems (1994)) in our context.

Definition 2.6. Let R be any ring that is isomorphic to a subring of the ring \mathcal{O} of entire functions with pointwise defined multiplication. Denote the embedding $R \rightarrow \mathcal{O}$ by \mathcal{L} . A finitely presented R -system with presentation matrix P is said to be spectrally controllable if the \mathcal{O} -matrix $\hat{P} = \mathcal{L}(P)$ satisfies

$$\exists k \in \mathbf{N} : \forall \sigma \in \mathbf{C} : \text{rk}_{\mathbf{C}} \hat{P}(\sigma) = k$$

The following proposition establishes a first algebraic characterisation for spectral controllability

Proposition 2.4. (See Woittennek and Mounier (2010)). Let R be any ring that is isomorphic to a subring of the ring \mathcal{O} of entire functions with pointwise defined multiplication. A finitely presented R -system is spectrally controllable if, and only if, the module $\Sigma_{\mathcal{O}} = \mathcal{O} \otimes_R \Sigma$ is torsion free.

The next proposition strongly links spectral controllability and R -freeness, which yields a fairly constructive character.

Proposition 2.5. (See Woittennek and Mounier (2010)). Let R be any Bézout domain that is isomorphic to a subring of \mathcal{O} with the embedding $R \rightarrow \mathcal{O}$ denoted by \mathcal{L} . Then the notions of spectral controllability and R -free controllability are equivalent if and only if \mathcal{L} maps non-units in R to non-units in \mathcal{O} .

2.3 Decomposition

The classical decomposition of a module over a PID into its torsion and torsion free parts generalizes to Bézout domains (see, e.g. (Hazewinkel, 1995, p. 381)).

Proposition 2.6. Let R be a Bézout domain, and Σ an R -module. One has the following decomposition:

$$\Sigma = t\Sigma \oplus \Sigma/t\Sigma$$

where $t\Sigma$ is the torsion submodule of Σ . \square

3. FINITE DIMENSIONAL AND DELAY SYSTEMS

3.1 Finite dimensional systems

Two common instances of R -systems are given when R is the ring of differential operators $\mathbf{R}[\frac{d}{dt}]$, which yields

continuous time linear finite-dimensional systems Fliess (1990), and when R is $\mathbf{R}[\delta_h]$, with δ_h the shift operator of delay h , h being the sampling interval, with δ_0 being the identity operator. This yields linear discrete time finite-dimensional systems Fliess (1992).

On such rings (being common examples of principal ideal rings), all the notions of torsion freeness, projectivity and freeness coincide, resulting in the well known Kalman controllability notion. See Fliess (1990); Fliess and Bourlès (1997) for several fruitful consequences of this framework.

3.2 Various rings and delay systems

We shall consider the following rings in the sequel:

- $\mathcal{R}_p = \mathbf{R}[\frac{d}{dt}, \delta_L]$, the polynomial ring; a ring on which delay systems are defined. In the incommensurate delay case, it becomes $\mathcal{R}_p = \mathbf{R}[\frac{d}{dt}, \delta_L]$, where $\delta_L = (\delta_{L_1}, \delta_{L_2}, \dots, \delta_{L_r})$ and the dimension of the \mathbf{Q} vector space $[L_1, \dots, L_r]_{\mathbf{Q}}$ is r .
- $\mathcal{R}_e = \mathbf{R}[\frac{d}{dt}, e^{-L\frac{d}{dt}}]$, the ring of exponential polynomials, where an analytic relationship between the ring indeterminates is emphasized. This ring is a subring of $\mathbf{R}[[\frac{d}{dt}]]$, the power series ring in $\frac{d}{dt}$. It is commutative, since $\frac{d}{dt}e^{-h\frac{d}{dt}} = e^{-h\frac{d}{dt}}\frac{d}{dt}$. In the incommensurate delay case, it becomes $\mathcal{R}_e = \mathbf{R}[\frac{d}{dt}, e^{-L\frac{d}{dt}}]$.
- $\mathcal{R}_a = \mathbf{R}[\frac{d}{dt}, e^{-L\frac{d}{dt}}, e^{L\frac{d}{dt}}]$, the ring of exponential delays and advances. In the incommensurate delay case, it becomes $\mathcal{R}_a = \mathbf{R}[\frac{d}{dt}, e^{-L\frac{d}{dt}}, e^{L\frac{d}{dt}}]$.
- $\mathcal{R}_d = \mathbf{R}(\frac{d}{dt})[e^{-L\frac{d}{dt}}, e^{L\frac{d}{dt}}] \cap \mathcal{O}$, the distributed delay ring, where \mathcal{O} denotes the ring of entire functions. This ring includes distributed delays. In the incommensurate delay case, it becomes $\mathcal{R}_d = \mathbf{R}(\frac{d}{dt})[e^{-L\frac{d}{dt}}, e^{L\frac{d}{dt}}] \cap \mathcal{O}$.
- $\mathcal{R}_s = \mathbf{R}[\delta_h]$, the discrete delay, a natural ring when considering purely discrete systems. This ring is isomorphic to $\mathcal{R}_s^e = \mathbf{R}[e^{-h\frac{d}{dt}}]$. In the incommensurate delay case, it becomes $\mathcal{R}_s = \mathbf{R}[\delta_h]$. The latter ring will not be used here, since in computer implementations, only commensurate delays are used.

3.3 Delay systems

The definition we shall adopt is the following.

Definition 3.1. A *pointwise linear delay system* is a finitely generated \mathcal{R}_p -module. A *pointwise linear commensurate delay system* is a finitely generated \mathcal{R}_p -module. \square

Definition 3.2. Let Λ be a pointwise linear delay system. We shall call the *exponential version* of Λ , denoted by Λ_e the \mathcal{R}_e -system obtained by the functor isomorphism ψ_e^* from $\mathbf{Mod}_{\mathcal{R}_p}$ to $\mathbf{Mod}_{\mathcal{R}_e}$ (where \mathbf{Mod}_R is the category of R -modules, for a ring R), with ψ_e the ring isomorphism:

$$\begin{aligned} \psi_e : \mathcal{R}_p &\longrightarrow \mathcal{R}_e \\ \delta_L &\longmapsto e^{-L\frac{d}{dt}} \end{aligned}$$

\square

Definition 3.3. Let Λ be a pointwise linear delay system. The extension of scalars $\mathcal{R}_a \otimes_{\mathcal{R}_e} \Lambda_e$ is called the *advanced extension*, or \mathcal{R}_a -extension of Λ , and is denoted as Λ_a . The extension of scalars $\mathcal{R}_d \otimes_{\mathcal{R}_e} \Lambda_e$ is called the *distributed extension*, or \mathcal{R}_d -extension of Λ , and is denoted as Λ_d . \square

4. DISCRETIZATION

We shall define the discretization in the following way.

Definition 4.1. Let Λ be a delay system, let h be a real number dividing L (i.e. such that $L = Nh$ with $N \in \mathbf{N}$). Consider then the ring morphism $\phi_{\mathbb{D}_{p,q,h}}$, where $p, q \in \mathcal{R}_s$

$$\begin{aligned} \phi_{\mathbb{D}_{p,q,h}} : \mathcal{R}_p &\rightarrow (q)^{-1}\mathcal{R}_s \\ \frac{d}{dt} &\mapsto \frac{p(\delta_h)}{q(\delta_h)} \end{aligned}$$

this morphism corresponding to an identity approximation, i.e., p and q being such that

$$\lim_{h \rightarrow 0} \frac{p(e^{-h\frac{d}{dt}})}{q(e^{-h\frac{d}{dt}})} = \frac{d}{dt}$$

A *discretizer* of Λ with sampling period h is a functor $\mathbb{D}_{p,q,h} : \mathbf{Mod}_{\mathcal{R}_p} \rightarrow \mathbf{Mod}_{(q)^{-1}\mathcal{R}_s}$ built in the following way:

$$\begin{aligned} \tilde{\Lambda} &= \mathbf{R}[\frac{d}{dt}, q(\delta_h)^{-1}] \otimes_{\mathcal{R}_p} \Lambda \quad (\text{extn}) \\ \mathbb{D}_{p,q,h}\Lambda &= \phi_{\mathbb{D}_{p,q,h}}^* \tilde{\Lambda} \quad (\text{rest}) \end{aligned}$$

where (extn) is an extension of scalars, (rest) a restriction of scalars, and $\phi_{\mathbb{D}_{p,q,h}}^*$ is the restriction of scalars functor associated to $\phi_{\mathbb{D}_{p,q,h}}$. \square

Remark 4.1. Let us remark that

$$\mathbf{R}[\delta_L] = \mathbf{R}[\delta_h^N] \subset \mathbf{R}[\delta_h] = \mathcal{R}_s$$

Let us give three examples of this general definition.

Examples 4.1. • **Backward difference discretization.** In this case, the morphism $\phi_{\mathbb{D}_{p,q,h}}$ is

$$\phi_{\mathbb{D}_{p,q,h}}(\frac{d}{dt}) = \frac{1 - \delta_h}{h}$$

i.e. $p = (1 - \delta_h)/h$, and $q = 1$. We shall denote $\mathbb{D}_{B,h}$ this discretization.

• **Forward difference discretization.** In this case, the morphism $\phi_{\mathbb{D}_{p,q,h}}$ is

$$\phi_{\mathbb{D}_{p,q,h}}(\frac{d}{dt}) = \frac{\delta_h^{-1} - 1}{h} = \frac{1 - \delta_h}{h\delta_h}$$

i.e. $p = (1 - \delta_h)$, and $q = h\delta_h$. We shall denote $\mathbb{D}_{F,h}$ this discretization.

• **Tustin discretization.** In this case, the morphism $\phi_{\mathbb{D}_{p,q,h}}$ is

$$\phi_{\mathbb{D}_{p,q,h}}(\frac{d}{dt}) = \frac{2(1 - \delta_h)}{h(1 + \delta_h)}$$

i.e. $p = 2(1 - \delta_h)$, and $q = h(1 + \delta_h)$. We shall denote $\mathbb{D}_{T,h}$ this discretization. \square

5. INTERPLAY BETWEEN CONTROLLABILITY AND SAMPLING

5.1 Ring and dimension level analysis

On a gross level, one has the following proposition, whose proof follows directly from the results in Section 2.

Proposition 5.1. Let Λ be a delay system. We have the following:

- (1) For Λ : \mathcal{R}_p -free $\Leftrightarrow \mathcal{R}_p$ -projective $\not\equiv \mathcal{R}_p$ -torsion free

- (2) For $\Lambda_e: \mathcal{R}_e$ -free $\Leftrightarrow \mathcal{R}_e$ -projective $\not\equiv \mathcal{R}_e$ -torsion free
- (3) For $\Lambda_{\mathcal{R}_d}: \mathcal{R}_d$ -free $\Leftrightarrow \mathcal{R}_d$ -projective $\Leftrightarrow \mathcal{R}_d$ -torsion free
- (4) For $\mathbb{D}_*\Lambda: \mathcal{R}_s$ -free $\Leftrightarrow \mathcal{R}_s$ -projective $\Leftrightarrow \mathcal{R}_s$ -torsion free,
 where $\mathbb{D}_* = \mathbb{D}_{p,q,h}$, with $p, q \in \mathcal{R}_s$.

Remark 5.1. Thus, from these general module properties implications and equivalences, one is tempted to conclude that from a controllability viewpoint, to work with $\Lambda_{\mathcal{R}_d}$ in a continuous setting or to work on $\mathbb{D}_*\Lambda$ in a discrete setting is equivalent. We shall see in the sequel that this is not true. \square

5.2 A Disturbing situation

On the ring and dimension level analysis, things appear to be easy to handle. But these levels are quite coarse, and when one looks deeper into a finer, Fitting ideal analysis, the situation appears to be more tricky, as the following examples highlight.

Example 5.1. Consider the following example

$$\dot{x}(t) = s_0x(t) + (1 - hs_0)^N u(t) - u(t - L)$$

whose associated module is Λ . The associated presentation matrix is

$$\begin{pmatrix} \frac{d}{dt} - s_0 & (1 - hs_0)^N - e^{-L\frac{d}{dt}} \end{pmatrix} = \begin{pmatrix} m_1(\frac{d}{dt}, e^{-L\frac{d}{dt}}) & m_2(\frac{d}{dt}, e^{-L\frac{d}{dt}}) \end{pmatrix}$$

Hence, the system is spectrally controllable if $s_0 \neq 0$ and not spectrally controllable if $s_0 = 0$. Through backward difference discretization, we get

$$\begin{pmatrix} \frac{1 - \delta_h}{h} - s_0 & (1 - hs_0)^N - \delta_h^N \end{pmatrix} = (1 - hs_0 - \delta_h) \begin{pmatrix} \frac{1}{h} & r(\delta_h) \end{pmatrix}$$

Thus $\mathbb{D}_{B,h}\Lambda_e$ is non controllable, whatever the value of s_0 is. \square

Example 5.2. Another example is

$$h\dot{x}(t) = x(t) + u(t - L)$$

whose associated presentation matrix is

$$\begin{pmatrix} 1 - h\frac{d}{dt} & e^{-L\frac{d}{dt}} \end{pmatrix}$$

and thus, the system is spectrally controllable. Its backward difference discretization has the presentation matrix

$$\begin{pmatrix} 1 - h\frac{1 - \delta_h}{h} & \delta_h^N \end{pmatrix} = (\delta_h \quad \delta_h^N)$$

Hence, the discretized system is non controllable. \square

Example 5.3. Let p and s_0 such that $p(s_0) = 0$. Consider the system

$$\dot{x}(t) = s_0x(t) + p(e^{-L\frac{d}{dt}})(u)(t)$$

The associated presentation matrix is

$$\begin{pmatrix} \frac{d}{dt} - s_0 & p(e^{-L\frac{d}{dt}}) \end{pmatrix}$$

and the system is not spectrally controllable. The discretization $\mathbb{D}_{p,q,h}$ has the following presentation matrix

$$\begin{pmatrix} \frac{p(\delta_h)}{q(\delta_h)} - s_0 & p(\delta_h)^N \end{pmatrix}$$

which has no common factor. Hence, the discretized system is controllable. \square

6. CONCLUSION

We have exhibited some discretizer notion of for delay systems envisioned through an algebraic module theoretic setting. A ring theoretic and dimension analysis has been conducted, under which strong similarities occur between properties of the continuous system on a distributed delay ring and of the discrete system on the difference operator ring. Several simple examples showed that this analysis is not sufficient and one needs to investigate further through finer tools, such as the Fitting ideal. This is the topic of further research, in progress.

ACKNOWLEDGMENTS

The authors thank an anonymous reviewer for most valuable and helpful comments that helped the authors to improve the overall quality of the paper.

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